Environment HSI TO v_{hsi} F_{hsi} v_{to} F_{to} passive Reduction/ Reconstr Network LDR Algorithms Frame-based Sample -based Interpolative Frequency -based Extrapolative Direct

Lossy Data Reduction Methods for Haptic Telepresence Systems

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Abstract— Telepresence systems are often deployed in scenarios Where communication bandwidth is limited. Consequently, data Frequency-based between operator and teleoperator has to be reduced. • Wavelet-based case of haptic telepresence, data reduction has an influence

• eves the stability of the overall system. This paper provides a • step towards a systematic framework for communication data bandwidth reduction in haptic telepresence systems discussing • Datability for a class of lossy data reduction (LDR) algorithms. Flexibility invaluation and experimental results validate the efficacy.

Passivity Endeavors RM

I. INTRODUCTION

Telepresence enables a human operator to perceive and maniptilate a remote environment. The human operator handles the human system interface (HSI) thereby commanding the robotic perception (TO) to perform actions in a remote environment. HSI and TO exchange command and sensor feedback signals Reconstr. over a communication network (COM). To achieve holistic refinote immersion, multiple modalities of human perception are addressed including the visual, auditory, and haptic sense, here Fig. 1 for an illustration. Applications for telepresence are $g_{rm,l}$



Fig. 1. Multimodal telepresence system.

focus of this paper is on haptic telepresence systems. Haptic perception involves both tactile perception through the skin, like vibrations, and kinesthetic perception of position, force, motion of joints and muscles [1]. By the haptic command and feedback signals energy is bilaterally exchanged between the HSI and the TO. Thereby a global control loop is closed via the communication network (COM). Main objectives in the control system design are stability and transparency; ideal transparency means that the operator should feel as if directly being present (immersed) in the remote environment. The key challenges associated with the loop closed over the COM are time delay and limited communication resources. While the time delay problem is well treated in the known literature [2]– [4], very few researchers in the area of telepresence consider communication resource limitations, e.g. [5]. Quantization to reduce data bandwidth in telepresence has been investigated by [6]–[9], however, stability (passivity) is not investigated. The lossless compression scheme proposed in [6] results in a tradeoff between compression efficiency and time delay required for compression. Differential pulse code modulation (DPCM) together with a fixed rate quantization has been proposed in [8], adaptive DPCM together with Huffman coding has been considered in [9].

In order to provide high immersion with respect to haptic perception, HSI and TO must provide high manipulability [10]. This means, both robots must have several degrees of freedom, resulting in several sensors and several actuators resulting in a large amount of communication data. Generally, the communication bandwidth over common purpose networks, such as the Internet, is limited. Severe communication constraints are induced in space or underwater applications. This motivates investigations towards algorithms for haptic data reduction. In order to design data reduction algorithms it is effective to discard irrelevant (unperceivable) information. The result is a lossy data reduction (LDR) algorithm. Any LDR algorithm also must guarantee stability (passivity) of the overall system. Main contribution of this paper is a classification scheme for a class of LDR algorithms. Sufficient stability (passivity) conditions are derived. Selected LDR algorithms derived from the framework are discussed in detail along with simulation and experimental results.

The remainder of this paper is organized as follows: In Section II a brief background is presented followed evaluation criteria and a classification scheme in Sections III and IV. Simulations and experimental results are presented in Sections V and VI, respectively.

II. BACKGROUND

This section introduces basic concepts to stabilize haptic telepresence systems and the formal definition of transparency.

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Direct

• iDS quency-based

Wavelet-based

- eDS
- CQ
- DBC

FlexibilityA. Passivity

PSfrag replacements represents a sufficient condition for passivity (1). In case of sivity Endeavors Passivity is an energy-based concept, which provides sufficient further ansmission delayed by the time delay T in the forward and

RMstability conditions for interconnected complex systems such Gperatore backward path, the power balance (4) can be rewritten as stas telepresence systems. A passive system does not generate^{COM} FFTenergy. An observable system with zero initially stored energy . Remote **IFFT**is passive, if

 $\int_0^t P_{in}(\tau) \mathrm{d}\tau \ge 0, \qquad \forall t \ge 0,$

PM

Interpol

Reconstr

 g_l where P_{in} represents the instantaneous power input. If the g_r subsystems of the telepresence system are passive, then overall h_lstability can be ensured. As illustrated in Fig. 2 a telep h_r resence system can be interpreted as a connection of twog_{rm,l} ports terminated by one-ports at each end. In a velocity-force $g_{pm,l}$ architecture [2], [3] velocity is commanded to the teleoperator I_{Ref} and force is fed back to the human. The power input is I_{iDS2}



Fig. 2. Haptic telepresence system with velocity-force architecturenterpolative Frequency

then defined as scalar product between effort (force) and flow-based (velocity) variables, where

$$P_{in} = \mathbf{v}^{\mathbf{T}} \mathbf{F} := v_{hsi} F_{hsi} - v_{to} F_{to}.$$

is the power entering the COM-two-port with v denoting the • iDS velocity and F the force; subscripts for HSI and TO. frequency based DR algorithms discard or compress information to reduce a passive telepresence system the subsystem consisting Wavene basefile amount of data that has to be stored, processed, and human operator handling the HSI has to be passive. A trained • ensated i.e. the data transmission with LDR is no longer operator is supposed to assure this. The one-port consisting of • Geal. The LDR acts within the closed control loop. Stathe TO working in the remote environment must be passive DBGlity has to be ensured for the haptic telepresence system as well.

of constant communication delay, the active COM can be passivated by the scattering transformation introduced in [2], [3]. The bijective transformation of velocity and force into scattering variables is given by

$$g_l = \frac{bv_{hsi} + F_{hsi}}{\sqrt{2b}}, \qquad h_l = \frac{bv_{hsi} - F_{hsi}}{\sqrt{2b}},$$
$$g_r = \frac{bv_{to} + F_{to}}{\sqrt{2b}}, \qquad h_r = \frac{bv_{to} - F_{to}}{\sqrt{2b}}, \qquad ($$

where q describes the incident wave and h the reflected wave. The tuning parameter b > 0 represents the wave impedance. port including the scattering transformation is

$$\int_0^t P_{in}(\tau) \mathrm{d}\tau = \frac{1}{2} \int_0^t (g_l^2 - g_r^2 + h_r^2 - h_l^2) \mathrm{d}\tau.$$

It is straightforward to show that

$$\int_{0}^{t} (g_{l}^{2} - g_{r}^{2}) \mathrm{d}\tau \ge 0 \quad \text{and} \quad \int_{0}^{t} (h_{r}^{2} - h_{l}^{2}) \mathrm{d}\tau \ge 0, \ \forall t, \quad (5)^{I_{e}}$$

$$\int_0^t P_{in}(\tau) \mathrm{d}\tau = \frac{1}{2} \int_{t-T}^t (g_l^2 + h_r^2) \mathrm{d}\tau \ge 0 \qquad \forall t,$$

Environments $g_r(t) = g_l(t-T)$ and $h_l(t) = h_r(t-T)$. The system is (1) H**B**assive, i.e stable, for arbitrarily large constant delay.

TO. Transparency and Immersion

A telepresence system is called transparent if the operator is not able to distinguish between direct interaction with an environment and interaction by telepresence. Hence, transparency sets the benchmark of the telepresence system to achieve ideal remote immersion; in [11] the equality of position and force is termed transparency. In this paper the equivalent transparency definition of equal velocities v and forces F at HSI and TO ____is_ used as follows

$$v_{to} \stackrel{!}{=} v_{hsi}$$
 and $F_{hsi} \stackrel{!}{=} F_{to}$ (6)

Samplesing this definition of transparency, the integral of the -basguadratic velocity and force error

$$I = \int_0^t \left[(v_{hsi} - v_{to})^2 + (F_{to} - F_{hsi})^2 \right] \mathrm{d}\tau.$$
 (7)

Extrapolative I indicates a transparency measure. High cost I indicates a pirect and of (6).

III. LDR EVALUATION CRITERIA

Flexibility. With LDR. This is the major difference to standard coding

A COM-two-port with time delay is not passive. as a gorithms in audio and video applications. Based on the Phassivity framework as introduced in the previous section a general architecture for haptic telepresence system with FEDR is proposed: The bidirectional communication channel iFfs encapsulated by two-ports that perform a passive signal

Preduction and signal reconstruction as shown in Fig. 3. Time Interpoled average of the data transmission over the Reconstruction with the LDR algorithm itself are passivated using 3) the scattering transformation. Sufficient conditions for the passivity of LDR and such for the overall stability of the haptic telepresence system are presented in Section IV.

With stability/passivity as a necessary property, transparency Using (1), (2) and (3), the energy balance of the COM two- g_{rm} and compression ratio are the main performance objectives $g_{pm,l}$



Fig. 3. COM with LDR for haptic telepresence systems.

PSfrag replacements

in the LDR design. The following criteria are proposed to manal conditions are specified in scattering variables, cf. (3). evaluate passive LDR algorithms: Operator The conditions are stated exemplarily for the forward path, 1. Transparency: A LDR algorithm is called transparent combined by however, analogously apply to the backward path. Additional the human operator does not perceive a difference betweensivedata-preserving coding strategies (Huffman coding, etc.) are an approach without LDR and with LDR. Transparency pinnote not discussed for conciseness but can be readily integrated this sense is generally difficult to evaluate without human Wserheninto the classification scheme as either frame- or sample-based studies as the human perception characteristics has to be taken_{HSI}versions.

into account. Sufficient conditions for transparency (6) cannot TOA. Frame-based Approach: Frequency-Based LDR

be satisfied by any LDR algorithm because at least some v_{hsi} Frequency based LDR algorithms transform the signal from data is lost. Accordingly, any of the reviewed transparency Frequency based LDR algorithms transform the signal from measures, such as (7), gives a relative value appropriate for v_{to} the sender into the frequency domain. A (large) number the comparison of LDR algorithms. An absolute statement on F_{to} of samples are gathered into frames, then e.g the Fourier transparency remains difficult because of additional time delaysive transform is applied, and the data reduction is achieved by and estimation errors introduced by an LDR algorithm. Reduction/transmitting only dominant frequency components of the spec-2. Compression ratio: The compression ratio is defined heconstr. trum. At the receiver site another transformation translates the the ratio of the amount of original data in the uncompressed signal back into the time domain, which is then used for the lossless case to the amount of reduced data, normalized to local control loop. This approach introduces (possibly large) additional time delay

the amount of reduced data. The compression ratio is given by c:1. It is desirable to have a high compression ratio indicated by a high number c >> 1. As it represents a average value a high compression ratio does not implicitely satisfy possible communication bandwidth constraints. Therefore, the maximum data rate, i.e. the minimum instantaneous compression ratio, that can be guaranteed by a LDR algorithm, is of high interest.

In general, the design is subject to a trade-off between achievable transparency and achievable compression ratio, hence the search for the optimal algorithm can be formulated as a multi-ips objective optimization problem. Accordingly, the optimal LDR interpolative LDR approximates the incoming signal in the algorithm among a set of algorithms

- achieves the maximum transparency level given a certainers
 achieves maximum compression ratio, or
 achieves maximum compression ratio given a certainers
 achieves maximum compression ratio given a certainers
- achieves maximum compression ratio given a certain p_{BC} reconstructs the signal using the (e.g. spline) parameter vector Flexibility r_{BC} reconstructs the signal using the (e.g. spline) parameter vector flexibility r_{BC} is illustrated

For a guaranteed minimum instantaneous compression Frateovors. In Fig. 5 the structure of the COM with LDR is illustrated. (limited communication bandwidth) the same argument ap-RM the frame length t_F has to fulfill plies. ST

FFT

iFFT

IV. CLASSIFICATION OF LDR APPROACHES

$$\int_{t_j+t_F+T}^{t_j+2t_F+T} g_r(t)^2 dt \le \int_{t=t_j}^{t_j+t_F} g_l(t)^2 dt,$$
(8)

The top-down classification scheme proposed in this paper $_{\mathsf{PM}}$ divides the LDR approaches into *frame-based* and *sample* with t_j the starting time of frame j. That means, the energy based strategies, see Fig. 4. In a frame-based approach a proach a proach of the interpolated signal over the frame length t_F has to be number of samples is gathered to a frame before the LDR $_{g_l}$ less or equal to the energy of the original signal. There are becomes active, i.e. it induces additional time delay into the g_r two factors influencing transparency: 1) the induced additional closed loop. Stabilizing control measures such as the scattering h_i time delay equal to the frame length t_F ; 2) The error resulting transformation are necessary to guarantee stability. Sample- hr based methods act on the single sample rather than on a frame $f_{m,l}$ without introducing additional time delay.

 q_{nm} Frame-based methods are subdivided into frequency-based_{Ref} LDR and interpolative LDR, while the sample-based methods LDR_{iDS2} are subdivided into extrapolative and direct methods. These four LDR strategy classes are introduced in the following I_{CQ} and discussed with respect to passivity, transparency, and_{eDS2} compression ratio. Passivity conditions are presented. A haptig telepresence system with the constant time delay T in the forward and the backward path is assumed. Consequently,

into the telepresence system, where it is known from [12] that

already a small increase in time delay significantly impairs

transparency; this is the reason why the frequency-based

approach is not further elaborated in this paper. The frame

length needs to be (quite) large to encode low frequency

components of the signals. An advantage of the approach is

that a constant data compression ratio can be achieved.

B. Frame-based Approach: Interpolative LDR



Fig. 4. Classification scheme of haptic LDR algorithms.



Fig. 5. COM for interpolative LDR

from the reduction of the time series in the frame to a parameter vector **p** of a signal shaping algorithm. Given a signal shaping algorithm, the higher the dimension of the parameter vector the lower is the error. Low frame length, i.e. low additional time delay, and high dimension of the parameter vector will lead to good transparency. However, a high compression ratio can only be achieved for a high frame length and a low dimensional parameter vector. The compression ratio per frame is c:1 with

$$c = \frac{k_F}{\dim(\mathbf{p})},\tag{9}$$

and k_F the number of samples in one frame. The data rate is constant, hence the instananeous compression ratio is equal to the compression ratio on average. Communication bandwidth constraints can be satisfied. The trade-off between transparency and compression ratio is adjustable according to the requirements.

C. Sample-based Approach: Extrapolative LDR

The extrapolative strategy deploys an estimation of the next k_{es} samples, called estimation horizon, at the sender side to achieve a reduction. An extrapolation algorithm estimates the future signal parameters are transmitted within the allowable passivity (energy) limits. No additional time delay is induced. The structure is the same as shown in Fig. 5. Instead of an interpolation an extrapolation is performed and the resulting parameter vector **p** is transmitted over the network. Every k_{eh} samples an estimation of the next k_{eh} samples is conducted. To assure a passive estimation according to (5) the energy of the estimated samples has to be smaller than the difference between the unshaped signal energy and the energy used to estimate the past samples. Formally, the following passivity constraint has to hold

$$\int_{0}^{t_{j}} (g_{l}^{2} - g_{r}^{2}) \mathrm{d}t \ge \int_{t_{j}+T}^{t_{j}+t_{eh}+T} g_{r}^{2} \mathrm{d}t, \qquad (10)$$

with t_i the time when a new estimation is performed and t_{es} the duration of the estimation horizon. Any estimation algorithm can be used as long as (10) is satisfied.

The advantages of the approach are similar to the advantages of the interpolative LDR. However, in place of an approximation this strategy uses an estimation and in place of frames it uses estimation series. The compression ratio is similar to (9) replacing k_F by k_{es} .

D. Sample-based Approach: Direct LDR

Within the direct LDR strategy, the reduction scheme is performed on each sample directly. An example for a direct

LDR strategy is coarser quantization. The structure is again shown in Fig. 5. Instead of an interpolation the direct scheme is deployed. The passivity condition (5) is fulfilled by assuring that the absolute value of the incoming sample $g_r(t+T)$ has to be decreased or left unchanged compared to the associated sent sample $q_l(t)$

$$|g_l(t)| \ge |g_r(t+T)|$$
. (11)

The main advantage of the approach is due to its direct character: No delays are induced and the remote immersion is left unchanged beside the additional quantization noise. Furthermore, the passivity constraint (11) is straightforward and easy to fulfill. The average data reduction as well as the upper bandwidth can be adjusted by the chosen resolution.

A related approach is deadband control as proposed in [5], which results in non-uniform sampling and is therefore somewhat different. Approaches using (adaptive) differential pulse code modulation are also direct LDR algorithms, see [8], [9].

E. Summary: LDR classification

The interpolative, extrapolative, and direct strategy form a classification framework for many possible LDR algorithms. The conditions (8), (10), (11) assure passivity of algorithms of the specific class.

V. SIMULATIONS

This section provides simulations of three algorithms. Each of the algorithms represents one of the explained classes of algorithms: interpolative LDR, extrapolative LDR and direct LDR. The strategies are applied to the scattering variables. In order to evaluate the influence of the LDR algorithm only, the time delay is set to T = 0. The remote environment is modeled by an impedance with parallel spring-damper behavior with the transfer function 1/(s+1). The human operator exerts a velocity step. Its reaction is modeled by an admittance with serial spring-damper dynamics with the transfer function s/(s+0.5). The dynamics of HSI and TO are assumed to be negligible, they are modeled as ideal transducers with unity transfer functions. The sample frequency is 1000 Hz. The velocity and force responses of the HSI and the TO are presented to demonstrate the effect of the different algorithms. The transparency criterion (7) is computed over the simulation horizon of 5s and normalized to the highest value. For the standard approach without LDR, the transparency criterion is $I_{Ref} = 0$ as expected; naturally, the compression ratio is 1 : 1.

A. Interpolative LDR: Passive Interpolative Down Sampling

For the interpolative LDR a passive interpolative down sampling (iDS) is applied, the parameter vector p contains the mean value over the frame length. It is straightforward to show that this strategy satisfies the passivity condition (8). The results for the interpolative LDR for frame-lengths of $k_F = 2$ and $k_F = 50$ are shown in Fig. 6 and Fig. 7, respectively. The transparency is decreased for the higher frame-length resulting from additional time delay and interpolation error:

PSfrag replacements

time [s]

higher frequencies are filtered out. The transparency criterion it is very similar to Fig. 6. The transparency decreases with Human values are $I_{iDS2} = 0.05$ and $I_{iDS50} = 1$, respectively. As the area of the stimation horizon as observable from the transparency Frequencies are $I_{iDS2} = 0.03$ and $I_{iDS0} = 1$, respectively. The operator values $I_{eDS2} = 0.03$ and $I_{eDS0} = 0.61$. In contrast produced dimension of the transmitted parameter vector is $dim(\mathbf{p}) = COM$ criterion values $I_{eDS2} = 0.03$ and $I_{eDS0} = 0.61$. In contrast Reflection of the transmitted parameter refer to the interpolative LDR no additional delay is introduced, Bassive to the interpolative LDR no additional delay is introduced, (equivalent to a data reduction of 50% and 98%), respectively, the decreased transparency results from the estimation error Network is achieved. Environment only. As in the previous case, the compression ratio is 2:1



 I_{eDS50}

Algo

avelet-



B. Extrapolative LDR: Passive Extrapolative Down Sampling g_{π}

Similar to the interpolative LDR, for the extrapolative $LDR_{R_{ef}}$ a passive extrapolative down sampling (eDS) is applied. $I_{I_{DS}}$ Here again a single value is transmitted parameter $vec_{i,DSS}$ tor: $dim(\mathbf{p}) = 1$. This value is either the most recent value q_{CC} measured, if (10) is satisfied or a value computed such that $(10)_{PDS}$ is satisfied. This strategy is passive as straightforward to show, r_{eDS} The result for an estimation horizon of $k_{es} = 50$ is depicted in Fig. 8, the result for $k_{es} = 2$ is not explicitly shown as

HSI^{and 50}: 1, respectively.

TOC. Direct LDR: Coarser Quantization

 v_{hsi} As a direct LDR scheme a coarser quantization (CQ) is F_{hsi} implemented: the standard quantization of 16 Bit is reduced v_{to} to 8 Bit quantization. The passivity condition (11) is satisfied F_{to} by transmitting the next lower (in an absolute sense, i.e. time [s] closer to zero) quantization value. The coarser quantization

= 2 achieves 2:1 of 8 Bit, as snown in Fig. 2, integers Reconstr. noise deteriorating transparency. The immersion value is $I_{CQ} = 0.34$ while the compression ratio is 2:1.

Agorithms The trade-off between transparency and compression ratio, stated in Section III is clearly observable. Comparing the Sample transparency for the compression ratio of 2:1 the passive rpolative extrapolative downsampling approach is superior over the -based 5 other two approaches.

VI. EXPERIMENTS AND EVALUATION

Direct performed using a two degree-of-freedom (2-DoF) haptic • telepresence system. The experimental set up is shown in • iDSFig. 10. It consists of two identical SCARA-robots with two -based degrees of freedom connected to a PC. The link angles are et-based measured by an incremental encoder, the torque applied to • eDS each link by strain gauges. The sensor data are processed • CO in the PC running unter RT/Linux. All control algorithms time [s] • DBC(HSI force control, teleoperator velocity control in the joint

low Passivity Endeavors Matlab/Simulink models with realtime code generated from ^{RM}them. The control loops operate at a sampling rate of 1000 Hz STrepresenting the communication rate without data reduction FFT algorithms. The time delay is set to T = 0 in order to evaluate iFFT the influence of the LDR algorithms only. For each algorithm PM10 runs are performed where the human operator is moving ^{interpol.} in free space in the time interval 0..5 s, pushes into a damping Beconstr.environment (foam) in the time interval 5..10 s, and finally g_l pushes twice against a stiff surface in the time interval 10..15 s. Coarser quantization achieves 2:1 compression ratio. Noise g_r In Fig. 11 the velocity and force measured at the inner links

 h_l of HSI and TO are depicted for the reference (no LDR) h_r experiment for an example run. Example velocity and force ^{*m*,*l*} responses for interpolative down sampling (iDS) with $k_F = 2$,



Fig. 10. Experimental 2-DoF haptic telepresence system.

Evene the standard deviation of 8,Bit (CQ) are shown in Fig. 12, Algorithm Fig. 13 and Fig. 14, respectively. The mean values and under the standard deviation over the performed 10 runs of the Waveler based the standard deviation over the performed 10 runs of the based based the standard deviation over the performed 10 runs of the based based the standard deviation over the performed 10 runs of the based based the standard deviation over the performed 10 runs of the based based the standard deviation over the performed 10 runs of the based with the standard deviation over the performed 10 runs of the based bas

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Network defined in Section III, human user studies are necessary being



 I_{eDS} % ivity Endeavo_{kl}Fig. 12. Interpolative down sampling with frame length $k_F = 2$ shows good R_{ll} transparency.





Fig. 14. Coarser quantization of 8Bit: Transparency deteriorated.

VII. CONCLUSION

This paper proposes a systematic classification framework for lossy data reduction (LDR) algorithms in haptic telepresence systems. Two groups of sample-based and framebased approaches include direct, interpolative, extrapolative,



Fig. 15. Transparency comparison: Mean value and standard deviation over 10 runs for the reference approach (no LDR) and considered LDR algorithms.

and frequency-based algorithms. Sufficient passivity (stability) conditions are stated as a design guideline for LDR algorithms applicable to haptic telepresence systems with time delay. Simulation and experimental results validate selected LDR algorithms in their performance in terms of transparency and data compression ratio.

Future investigations are to analyze issues of human perception such as LDR algorithm design using psychophysical insight, e.g. just-noticable-difference thresholds for motion and forces.

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