

Bandwidth Efficient and Rate-Matched Low-Density Parity-Check Coded Modulation

Georg Böcherer, Patrick Schulte, Fabian Steiner

Chair for Communications Engineering
Technische Universität München
`patrick.schulte@tum.de`

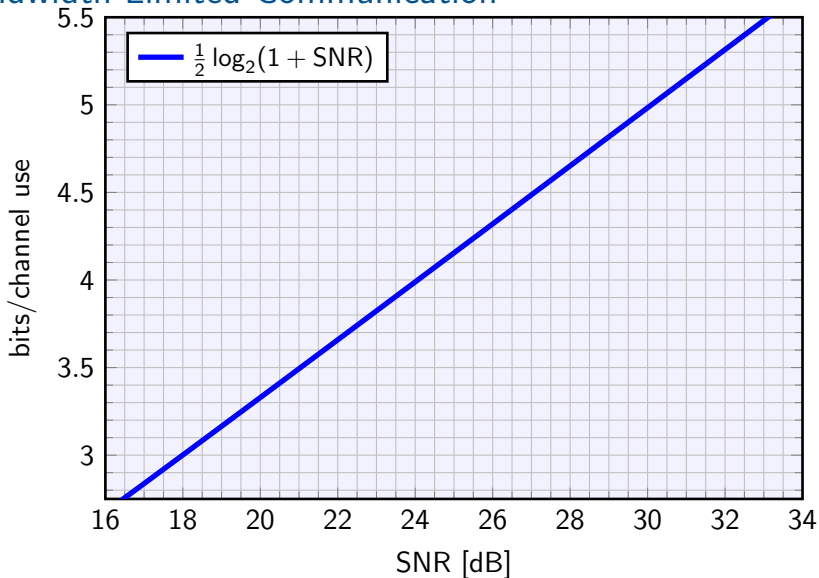
April 29, 2015
ITG Fachgruppentreffen Angewandte Informationstheorie

Outline

- Motivation: Bandwidth-Limited Communication
- Contribution: A New Coded Modulation Scheme
- Enabling Technology: Fixed-Length Distribution Matching

Bandwidth-Limited Communication

Bandwidth-Limited Communication



Higher Order-Modulation

- **Equidistant 2^m -ASK constellation:**

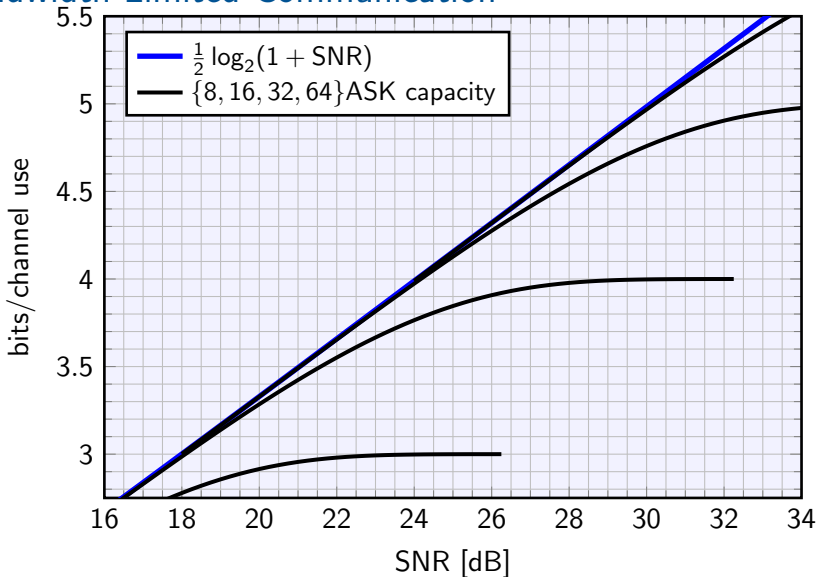
$$\mathcal{X} = \{\pm 1, \pm 3, \dots, \pm(2^m - 1)\}.$$

- **I/O-relation:**

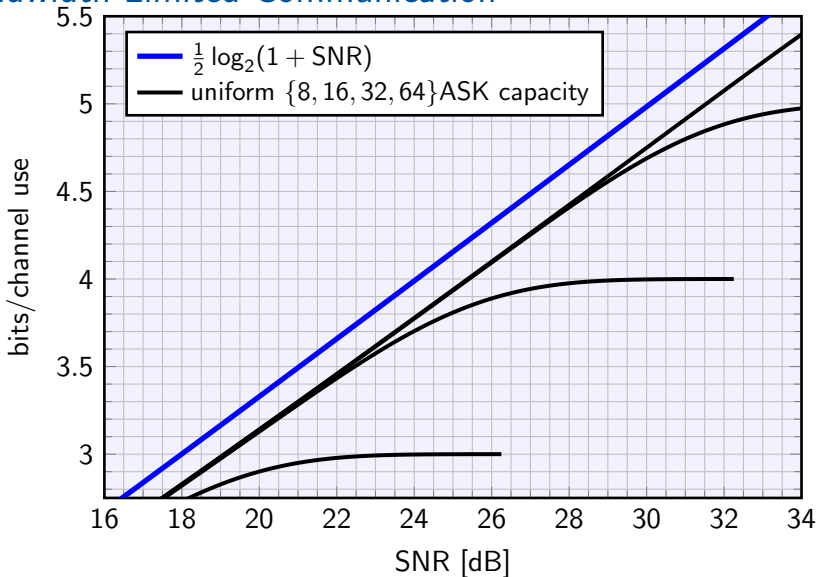
$$Y = \Delta \cdot X + Z.$$

- Noise Z is zero mean, variance one.
- Input X with distribution P_X on \mathcal{X} .
- Δ scales the constellation \mathcal{X} .

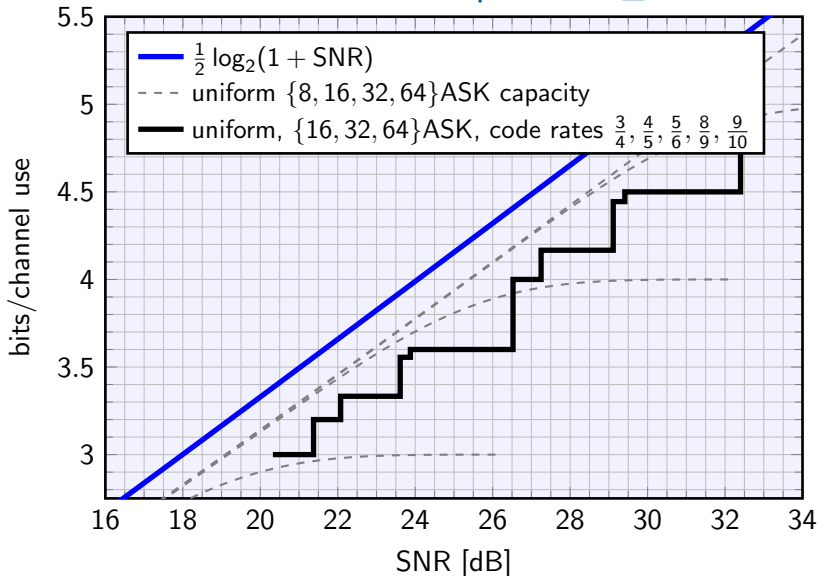
Bandwidth-Limited Communication



Bandwidth-Limited Communication



DVB-S2 LDPC Codes, Uniform Input, $FER \leq 10^{-3}$



Probabilistic Amplitude Shaping

ASK Capacity

- **ASK capacity:**

$$C_{\text{ask}}(P) = \max_{\Delta, P_X: E[|\Delta X|^2] \leq P} I(X; \Delta X + Z).$$

- Capacity-achieving distribution P_{X^*} .

⇒ Construct a codebook with entries $\sim P_{X^*}$.

Capacity-Achieving Distribution P_{X^*}

- P_{X^*} is **symmetric**:

$$P_{X^*}(x) = P_{X^*}(-x).$$

- P_{X^*} **factorizes**:

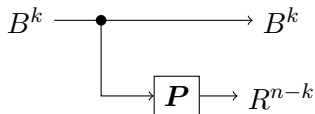
$$P_{X^*}(x) = P_S(\text{sign}(x)) \cdot P_A(|x|)$$

where $A := |X|$ and $S := \text{sign}(X)$.

- The sign S is **uniform**:

$$P_S(-1) = P_S(1) = \frac{1}{2}.$$

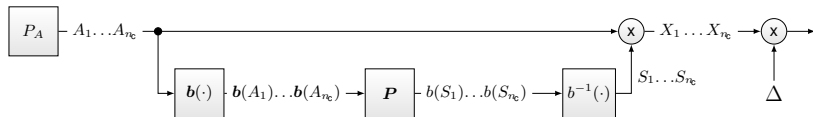
Shaping and Error Correction



- Binary systematic $k \times n$ generator matrix $\mathbf{G} = [\mathbf{I} | \mathbf{P}]$.
- Binary information B^k arbitrarily distributed.
- Uniform check bit assumption:

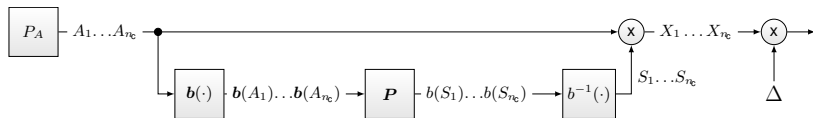
$$R_1, R_2, \dots \text{ iid } \sim \text{Bernoulli}(1/2).$$

Probabilistic Amplitude Shaping



- n_c number of channel uses.
- Represent amplitudes by $(m - 1)$ bits: $A \mapsto \mathbf{b}(A)$.
- Represent signs by 1 bit: $S \mapsto b(S)$.
- Multiply amplitude labels by parity matrix \mathbf{P} to get sign labels.
- Code rate $c = (m - 1)/m$.
- $P_{X_i} = P_{X^*}!$

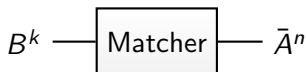
Rate Compatibility



- Control rate $H(A)$ via P_A .
- Control power $E[|\Delta X|^2]$ via Δ .

Enabling Technology: Distribution Matching

Emulating discrete memoryless source P_A :



- B_i iid Bernoulli(1/2).
- \bar{A}_i approximately iid $\sim P_A$ in the sense of

$$\frac{D(P_{\bar{A}^n} \| P_A^n)}{n} \xrightarrow{n \rightarrow \infty} 0, \quad \frac{k}{n} \xrightarrow{n \rightarrow \infty} H(A)$$

- invertible: B^k can be recovered from \bar{A}^n with zero error.

Constant Composition Codes

The empirical distribution of a output sequence \mathbf{c} of length n is

$$P_{\mathbf{A},\mathbf{c}}(a) := \frac{n_a(\mathbf{c})}{n},$$

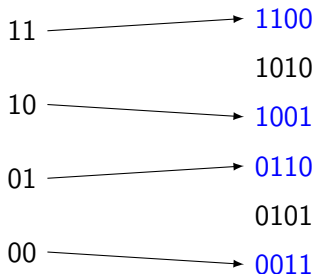
- $n_a(\mathbf{c}) = |\{i : c_i = a\}|$ number of occurrences of symbol a in \mathbf{c}
- $P_{\mathbf{A},\mathbf{c}}$ is the *type* of \mathbf{c}
- $\mathcal{T}_{P_{\mathbf{A}}}^n$ is set of all n -type $P_{\mathbf{A}}$ sequences
- Codebook $\mathcal{C}_{\text{ccdm}} \subseteq \mathcal{A}^n$ is called a *constant composition code* if all codewords are of the same type, i.e. $\mathcal{C}_{\text{ccdm}} \subseteq \mathcal{T}_{P_{\mathbf{A}}}^n$

Constant Composition Distribution Matching

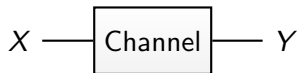
Idea

Use codewords of the same type

- Choose sequence length of the matcher output n
- Choose n -type approximation $P_{\bar{A}}$ of the target distribution P_A
- Choose $m = \lceil \log_2 |\mathcal{T}_{P_{\bar{A}}}^n| \rceil$
- Construct a unique mapping $\{0, 1\}^m \rightarrow \mathcal{T}_{P_{\bar{A}}}^n$



Bit-Metric Decoding (1)



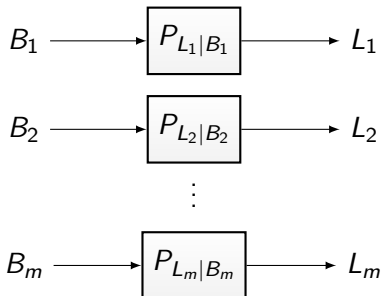
- $\text{label}(X) = b(S)\mathbf{b}(A) = B_1 \cdots B_m$.
- Demapper soft-decision:

$$L_i = \log \frac{P_{B_i|Y}(0|Y)}{P_{B_i|Y}(1|Y)}, \quad i = 1, \dots, m.$$

- No iterative demapping!

Bit-Metric Decoding (2)

- To the receiver, the channel appears as



Achievable Rate (2)

- Achievable rate:¹

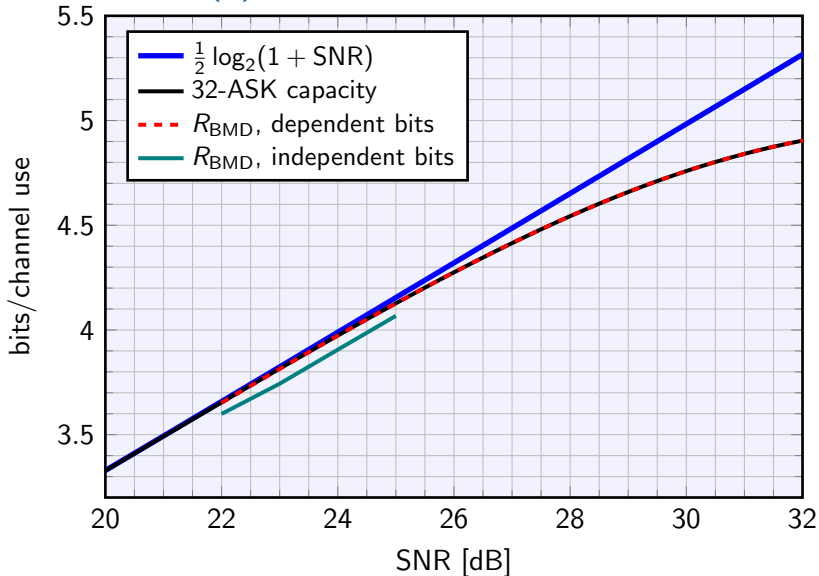
$$\begin{aligned}
 R_{\text{BMD}} &= H(\mathbf{B}) - \sum_{i=1}^m H(B_i | L_i) \\
 &= \underbrace{\sum_{i=1}^m I(B_i; L_i)}_{(*)} - D(P_{\mathbf{B}} \| \prod_i P_{B_i}).
 \end{aligned}$$

- (*) known as “BICM capacity”, “pragmatic capacity”, ...²

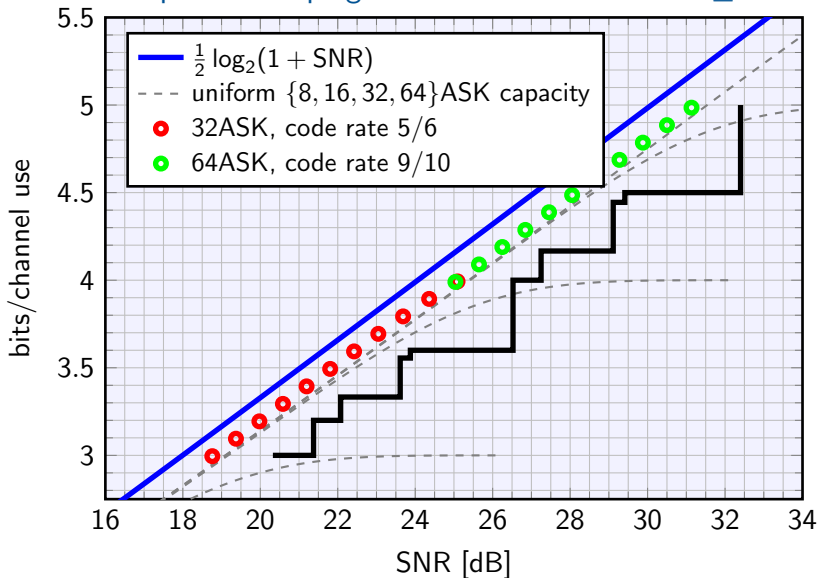
¹G Böhcherer [Achievable Rates for Shaped Bit-Metric Decoding](http://arxiv.org/abs/1410.8075), <http://arxiv.org/abs/1410.8075>.

²Guillen i Fabregas, Martinez [Bit-interleaved coded modulation with shaping](#), ITW 2010.

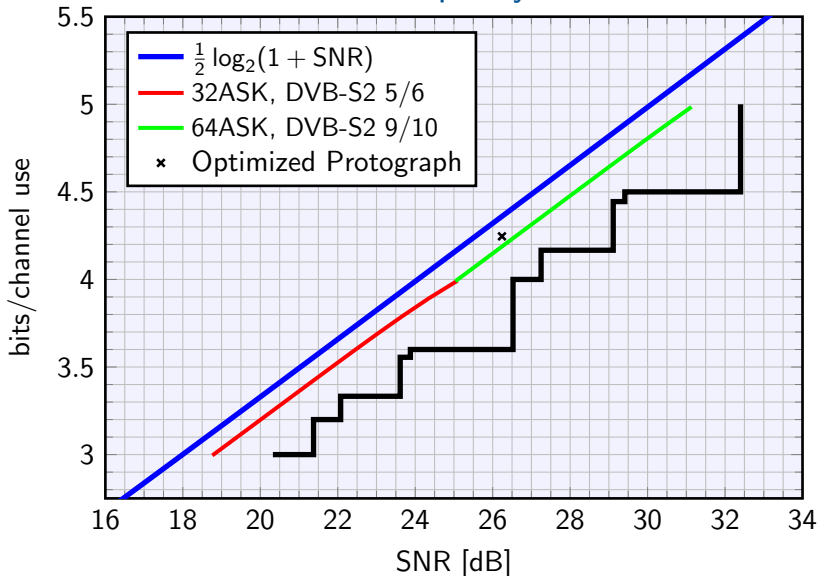
Achievable Rate (3)



Probabilistic Amplitude Shaping with DVB-S2 Codes $\text{FER} \leq 10^{-3}$



FER 10^{-3} within 0.69 dB of Capacity



References

- G Böcherer, P. Schulte, F. Steiner **Bandwidth Efficient and Rate-Matching Low-Density Parity-Check Coded Modulation**, <http://arxiv.org/abs/1502.02733>.
- P.Schulte, G Böcherer: **Constant Composition Distribution Matching**, <http://arxiv.org/abs/1503.05133>.
- G Böcherer **Achievable Rates for Shaped Bit-Metric Decoding**, <http://arxiv.org/abs/1410.8075>.
- F. Steiner, G Böcherer, G. Liva **Protograph-Based LDPC Code Design for Bit-Metric Decoding**, ISIT 2015.